

UNITED STATES PATENT APPLICATION

FOR

METHOD AND APPARATUS FOR DISSIPATIVE CLAMPING OF AN
ELECTRICAL CIRCUIT

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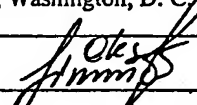
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METHOD AND APPARATUS FOR DISSIPATIVE CLAMPING OF AN ELECTRICAL CIRCUIT

BACKGROUND OF THE INVENTION

5 Field of the Invention

The present invention relates generally to electrical circuits and, more specifically, the present invention relates to electrical circuit clamping.

Background Information

Electronic devices use power to operate. Switched mode power supplies
10 are commonly used due to their high efficiency and good output regulation to power many of today's electronic devices. In a known switched mode power supply, a low frequency (e.g. 50 or 60 Hz mains frequency), high voltage alternating current (AC) is converted to high frequency (e.g. 30 to 300 kHz) AC, using a switched mode power supply control circuit. This high frequency, high
15 voltage AC is applied to a transformer to transform the voltage, usually to a lower voltage, and to provide safety isolation. The output of the transformer is rectified to provide a regulated direct current (DC) output, which may be used to power an electronic device. The switched mode power supply control circuit usually provides output regulation by sensing the output and controlling it in a closed
20 loop.

To illustrate, Figure 1 is a schematic of a known forward power converter
101. A switch Q1 103 turns on and off in response to a control 105 to provide a regulated DC output voltage V_{OUT} 129 from an unregulated DC input voltage V_{IN} 127. In one embodiment, control 105 and switch Q1 103 are included in a

switching regulator, which may be used to regulate the output voltage V_{OUT} 129 .

This topology is well known and its operation is well documented.

Every forward converter must have a way to set the voltage on the primary winding 107 of the transformer 109 during the time when the switch Q1 103 is
 5 off. A popular way to set the voltage is with a clamp network 111 connected across the primary winding 107. The known clamp network 111 illustrated in Figure 1 includes a resistor 113, a capacitor 115 and a diode 117 and absorbs and dissipates parasitic energy from the transformer 109 that is not delivered to the load 119 nor returned to the input 121. The balance of energy into the clamp
 10 network 111 through diode 117 and energy dissipated in 113 determines a clamp voltage V_{CLAMP} 123 that is necessary prevent saturation of the transformer 109.

Figure 2 shows with idealized waveforms how the voltage V_{SWITCH} 125 on switch Q1 103 is related to the input voltage V_{IN} 127 and the clamp voltage V_{CLAMP} 123. The clamp voltage V_{CLAMP} 123 must be high enough to prevent
 15 saturation of the transformer 109, but low enough to keep the voltage V_{SWITCH} 125 below the breakdown voltage of switch Q1 103.

Figure 3 shows the relationship between V_{CLAMP} 123 and V_{IN} 127 in a known power supply. As the input voltage V_{IN} 127 changes, the clamp voltage V_{CLAMP} 123 must be confined between the two boundaries shown in Figure 3. The
 20 maximum voltage boundary is a straight line determined by the breakdown voltage of switch Q1 103. The minimum voltage boundary is a curved line determined by the voltage necessary to keep the transformer 109 from saturation.

Figure 3 shows how the clamp voltage V_{CLAMP} 123 behaves with an RCD network, such as that illustrated in clamp network 111 of Figure 1. When the power converter 101 operates in continuous conduction mode, the clamp voltage V_{CLAMP} 123 stays substantially constant in response to changes in V_{IN} 127 at given load. The presence of leakage inductance in the transformer 109 causes the clamp voltage V_{CLAMP} 123 to change with load 119. It is higher for greater current and lower for less current. The result is a restricted range of permissible input voltage V_{IN} 127 that is shown in the shaded region of Figure 3.

SUMMARY OF THE INVENTION

Dissipative clamping methods and apparatuses are disclosed. In one aspect of the invention, a method includes switching a power supply input on an energy transfer element, regulating a power supply output by switching the power supply input on the energy transfer element, clamping a voltage on the energy transfer element to a clamp voltage and varying the clamp voltage in response to the power supply input. Additional features and benefits of the present invention will become apparent from the detailed description, figures and claims set forth below.

BRIEF DESCRIPTION OF THE DRAWINGS

The present invention detailed illustrated by way of example and not limitation in the accompanying figures.

Figure 1 is a schematic diagram illustrating a known forward converter
5 power supply.

Figure 2 is a timing diagram illustrating how the voltage on the switch is related to the input voltage and the clamp voltage in a known power supply.

Figure 3 is a diagram illustrating the relationship between the clamp voltage and the input voltage in a known power supply

10 Figure 4 is a block diagram illustrating one embodiment of the general elements of a dissipative clamp network in accordance with the teachings of the present invention.

Figure 5 is a diagram illustrating one embodiment of the relationship between the clamp voltage and the input voltage in accordance with the teachings
15 of the present invention.

Figure 6 is a schematic diagram illustrating one embodiment of a power supply using a dissipative clamp network in accordance with the teachings of the present invention.

DETAILED DESCRIPTION

Embodiments of methods and apparatuses for dissipatively clamping an electrical circuit such as a power supply regulator are disclosed. In the following description, numerous specific details are set forth in order to provide a thorough understanding of the present invention. It will be apparent, however, to one having ordinary skill in the art that the specific detail need not be employed to practice the present invention. In other instances, well-known materials or methods have not been described in detail in order to avoid obscuring the present invention.

Reference throughout this specification to "one embodiment" or "an embodiment" means that a particular feature, structure or characteristic described in connection with the embodiment is included in at least one embodiment of the present invention. Thus, the appearances of the phrases "in one embodiment" or "in an embodiment" in various places throughout this specification are not necessarily all referring to the same embodiment. Furthermore, the particular features, structures or characteristics may be combined in any suitable manner in one or more embodiments.

As an overview, Figure 4 shows the general elements of one embodiment of a dissipative clamp network 411 in an electrical circuit, such as for example a power supply 401, in accordance with the teachings of the present invention. As shown, an input voltage V_{IN} 427 is received at an input 421. A clamp network 411 is used to clamp the voltage V_{CLAMP} 423 across the primary winding 407 of a transformer 409. A switch 403 is coupled to primary winding 407 to drive

primary winding 407 in response to a control circuit (not shown). It is appreciated that transformer 409 is an inductive element and may be referred to as an energy transfer element or the like. A clamp diode D_{CLAMP} 437 provides a unidirectional path for the energy from the primary winding 407 of the transformer 409 to enter the clamp network 411. The energy is held by an energy storage element 435 and is lost through a dissipative element 433. In one embodiment, the dissipative element 433 is programmed by a signal S_1 439 from a sensing network 431. The sensing network 431 produces the programming signal S_1 439 from measurements of the input voltage V_{IN} 427, the voltage on the energy storage element 435 and a reference voltage V_{REF} 441 received by the sensing network 431. Thus, in one embodiment, energy stored in the leakage inductance of transformer 409 is dissipated in response to the input voltage V_{IN} 427.

In one embodiment, the dissipative element 433 is adapted in accordance with the teachings of the present invention, which can be viewed as having the effect of changing the value of the resistor 113 in the RCD clamp network 111 of Figure 1. The control from programming signal S_1 439 from sensing network 431 adjusts the energy balance to maintain a desired locus of clamp voltage over an extended range of input voltage as illustrated in Figure 5. As shown, in one embodiment the clamp voltage V_{CLAMP} 423 is varied substantially inversely linearly with respect to the input voltage V_{IN} 427 in accordance with the teachings of the present invention. Thus, in one embodiment, the clamp voltage V_{CLAMP} 423 is varied substantially independent of the power supply output and/or leakage inductance of transformer 409.

With the variation in clamp voltage V_{CLAMP} 423 as shown, the range of input voltages for V_{IN} 427 is increased in accordance with the teachings of the present invention. Indeed, various embodiments of the present invention allow operation over an extended range of input voltage for V_{IN} 427 while maintaining the clamp voltage V_{CLAMP} 423 at a high value within the minimum and maximum boundaries as shown in Figure 5. The higher voltages made possible by a variable clamp voltage V_{CLAMP} 423, such as illustrated in Figure 5, allows the use of parasitic capacitance in the primary winding 407 and secondary windings 443 to process some of the energy that otherwise would be dissipated in the clamp circuit 411.

Figure 6 is a schematic illustrating one embodiment of an electrical circuit such as for example a power supply 601 utilizing a dissipative clamp network 611 in accordance with the teachings of the present invention. As shown in the depicted embodiment, diode D3 637 provides the unidirectional path for energy from the primary winding 607 of the transformer 609 to enter the network 611 and capacitor C2 635 is the energy storage element of the clamp network 611. Zener diode VR1 645 and capacitor C3 647 make a stable reference voltage V_{REF} 641. In one embodiment, an N-channel metal oxide semiconductor (MOS) transistor Q2 is the principal dissipative element 633. In another embodiment, it is appreciated that other types of dissipative elements could be used in place of an N-channel MOS transistor such as for example p-channel MOS transistor, a bipolar transistor or the like or other future arising technology performing the function. In one embodiment, the sensing network in power supply 601 includes

the connection of resistors R1 649, R2 651, R3 653 and R4 655 with transistor Q3 657. The voltage on the gate of transistor Q2 633 is the programming signal S₁ 639 that adapts the dissipation to achieve the desired characteristic of operation.

In one embodiment, resistors R2 651 and R4 655 form a voltage divider that applies a scaled value of the sum of the input voltage V_{IN} 627 received at input 621 and the reference voltage V_{REF} 641 from Zener diode VR1 645 to the base of transistor Q3 657. The current flowing through R3 653 is proportional to the difference in voltage between the base of transistor 657 Q3 and the input voltage V_{IN} 627. The result is a current in the collector of transistor Q3 657 that decreases substantially linearly with increasing input voltage V_{IN} 627. The collector current in transistor Q3 657 produces a voltage drop through resistor R1 649 such that the voltage, or programming signal S₁ 639, on the gate of transistor Q2 633 is proportional to the weighted sum of the clamp voltage V_{CLAMP} 623 and the input voltage V_{IN} 627. The gate voltage on the gate of transistor Q2 633 controls the current in the dissipative element transistor Q2 633 to adjust the clamp voltage V_{CLAMP} 623 at a desired value for a given V_{IN} 627.

A first order analysis using reasonable engineering approximations reveals that the behavior of the circuit of power supply 601 is described by the expression

$$V_{CLAMP} = V_{REF} \left(1 + \frac{R1 \cdot R4}{R3(R2 + R4)} \right) - V_{IN} \left(2 - \frac{R1 \cdot R2}{R3(R2 + R4)} \right)$$

that describes a substantially straight line on the graph of V_{CLAMP} versus V_{IN}, as shown in Figure 5. An engineer can select values for resistances R1 649, R2 651,

R3 653 and R4 655 along with V_{REF} 641 to achieve the locus of desired operation as illustrated in Figure 5.

Figure 7 is a schematic illustrating one embodiment of an electrical circuit such as for example a power supply 701 utilizing a dissipative clamp network 711 in accordance with the teachings of the present invention. As shown in the depicted embodiment, diode D3 737 provides the unidirectional path for energy from the primary winding 707 of the transformer 709 to enter the network 711. Zener diode VR1 745 makes a stable reference voltage V_{REF} 741 relative to the circuit input negative rail of input 721. In one embodiment, a bipolar PNP transistor Q2 733 is the principal dissipative element. In another embodiment, it is appreciated that other types of dissipative elements could be used in place of a bipolar PNP transistor 733 such as for example a P channel MOSFET transistor. Resistor R1 753 is an optional additional dissipative element allowing the dissipated energy to be split between the bipolar transistor 733 and resistor R1 753. The energy is held by an energy storage element capacitor 735 and is lost through a dissipative elements transistor 733 and resistor 753.

In operation, the sum of the voltages V_{IN} 727 across the input 721 and V_{CLAMP} 723 across capacitor 735 is substantially constant. Thus, when V_{IN} 727 is relatively low, V_{CLAMP} 723 is relatively high. Conversely, when V_{IN} 727 is relatively high, V_{CLAMP} 723 is relatively low. Accordingly, V_{CLAMP} 723 is responsive to V_{IN} 727 received at input 721. Since the reference voltage V_{REF} 741 provided by zener diode VR1 745 is relative to the circuit input 721 negative rail, the operation of the clamp network 711 shown in Figure 7 provides a clamp that

limits V_{CLAMP} 723 across capacitor 735 to the locus of desired operation shown in Figure 5. In another embodiment it is appreciated that zener diode VR1 745 reference voltage V_{REF} 741 could be achieved with several lower voltage zener diodes in series.

5 It is appreciated that in the illustrated embodiment, transistor 733 in combination with resistor 753 and diode 745 embody a sensing network to sense V_{IN} 727 and thereby regulate the voltage across capacitor 735 such that the sum of V_{IN} 727 and V_{CLAMP} 723 remain substantially constant during circuit operation. Accordingly, the voltage V_{SWITCH} 725 across power switch Q1 703 is maintained
10 below a voltage limit of power switch Q1 703 in accordance with the teachings of the present invention.

 In the foregoing detailed description, the method and apparatus of the present invention has been described with reference to specific exemplary embodiments thereof. It will, however, be evident that various modifications and
15 changes may be made thereto without departing from the broader spirit and scope of the present invention. The present specification and figures are accordingly to be regarded as illustrative rather than restrictive.